



Outline – Frequency domain

- ☑ Introduction
- ☑ Fourier Transform
- ☑ Sampling Theory
- ☑ Systems
- ☑ Filters
- ☑ Windows
- Spectral Analysis



Parseval's theorem



The Fourier transform is unitary \rightarrow the sum (or integral) of the square of a function is equal to the sum (or integral) of the square of its transform.



 $\int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |X(f)|^2 df$ $\sum_{n=0}^{N-1} |x[n]|^2 = \sum_{k=0}^{N-1} |X[k]|^2$ Continuous Fourier transform



Discrete Fourier transform

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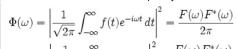


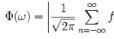
Signal energy

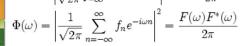


■ The energy spectral density describes how the energy (or variance) of a signal or a time series is distributed with frequency.









(only for finite energy signals)



Signal Power









$$\hat{S}[f] = |DFT\{x[n]\}|^2 = \left|\frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{\left(\frac{-f2\pi nf}{N}\right)^2}\right|^2$$

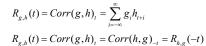


Correlation & Convolution











Convolution







Wiener-Khinchin theorem



■ The power spectrum is the Fourier transform of the auto-correlation function



 \blacksquare Power spectrum $S(f) = \int_{-\infty}^{\infty} R(\tau) e^{-j2\pi f \tau} \, d\tau$



 \blacksquare Autocorrelation $R(\tau) = \int_{-\infty}^{\infty} S(f) e^{j2\pi f \tau} \, df$



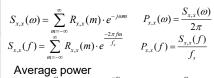
Power spectral density



Amount of power per unit (density) of frequency (spectral) as a function of the frequency



Power Spectrum



$$P_{x,x}(f) = \frac{2\pi}{f}$$



 $\begin{array}{l} \text{Average power} \\ \overline{P}(\omega_{\!\!\!1},\omega_{\!\!\!2}) = \int\limits_{\omega_{\!\!\!1}}^{\omega_{\!\!\!2}} P_{x,x}(\omega) d\omega - \int\limits_{\omega_{\!\!\!2}}^{\omega_{\!\!\!2}} P_{x,x}(\omega) d\omega = 2 \cdot \int\limits_{\omega_{\!\!\!2}}^{\omega_{\!\!\!2}} P_{x,x}(\omega) d\omega \end{array}$



Spectrum Estimation - Problems



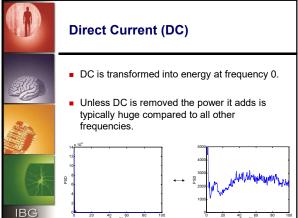
Leakage problems and side lobes.

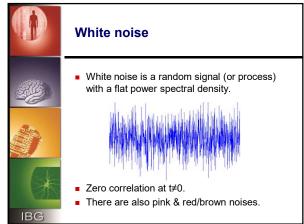


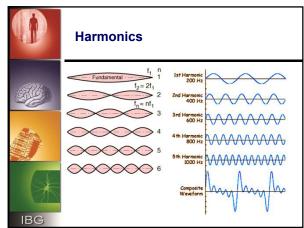
 Increased length of signal leads to increase in number of discrete frequency but not to increased accuracy at each frequency.

 Spectrum Estimation Methods
■ Non parametric estimators
 ■ Correlogram estimators
 ■ Parametric estimators
 Subspace estimators
 Example – Welch estimator
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 Things we find in the spectrum:
■ DC
 ■ White noise
 Harmonics
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with a flat power Zero correlation There are also pi	 — Direct Curren
White noise is a with a flat power • Zero correlation: • There are also pi	Unless DC is rentypically huge cofrequencies.
White noise is a with a flat power Zero correlation There are also pi	1BG 0 29 40 60 60
With a flat power Zero correlation: There are also pi	 White noise
Harmonics Fundamental 1, 1, 1, 1, 1, 1, 1, 1, 1, 1, 1, 1, 1, 1	■ White noise is a with a flat power
Harmonics Fundamental In-ent- In-ent- In-ent- In-ent- In-ent- In-ent- I	Zero correlation a
Fundamental I ₂ = 2I ₁ I _n = nI ₁ S	
	 — Harmonics
	$f_2 = 2t_1$ $f_n = nt_1$







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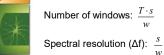
Do we have harmonies?

- Many biological processes lead to the formation of harmonies.
- Any square wave (for example a sudden change in a parameter) is transformed to multiple harmonies.
- The multiple frequencies may describe the same underlying process.



Temporal & spectral resolution

- Using windowed estimation (Welch/Bartlett) leads to a temporal / spectral resolution tradeoff.
- For a recording of *T* seconds sampled at s samples/sec and assessed using a w sample window:





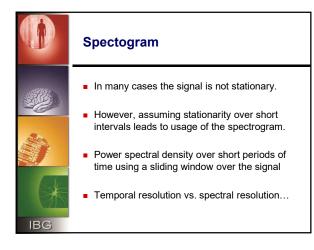


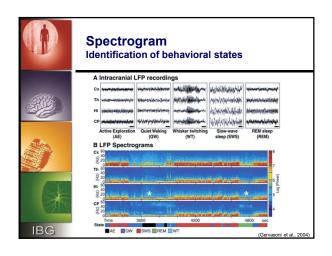
normalization of the signal.

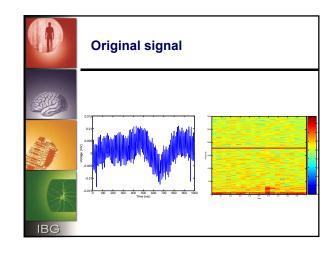
■ The absolute power depends heavily on the

- The relative power enables detecting the statistics of the signal at unfiltered frequencies.

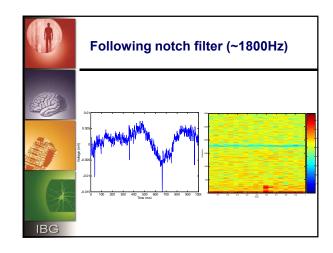
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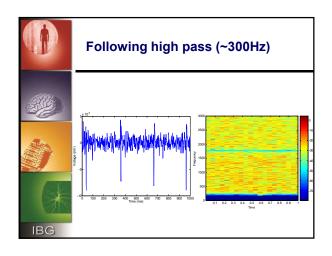


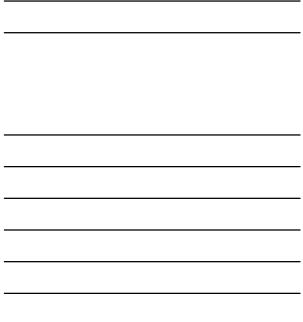


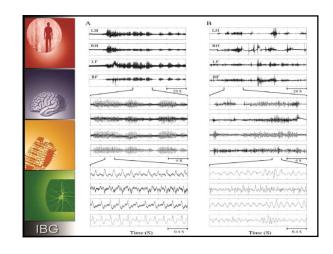




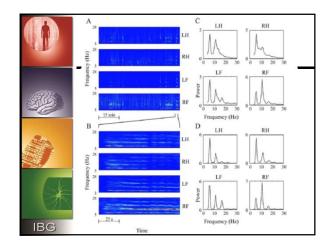


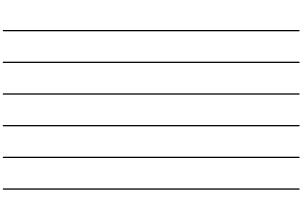


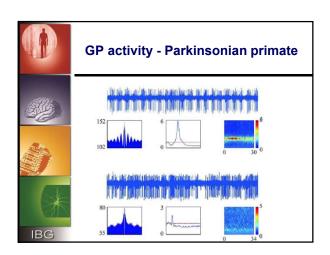


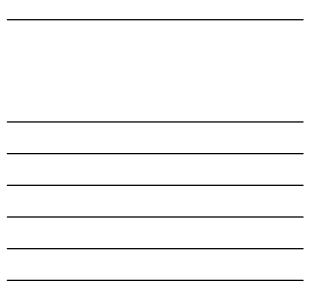














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Cross spectrum I

Multiplication of the Fourier transforms of the signals.

$$\hat{S}_{y,x}(\omega) = \frac{1}{N} Y(\omega) X^*(\omega)$$

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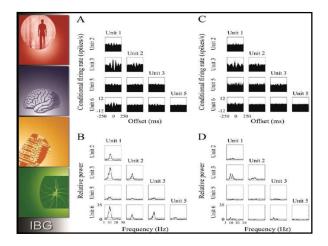
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Cross spectrum II

 Cross Spectral Density – defined as the transform of the cross correlation

$$S_{x,y}(\omega) = \sum_{m=-\infty}^{\infty} R_{x,y}(m) \cdot e^{-j\omega m}$$

 Two unrelated signals sharing common frequencies will have significant cross spectral density over finite length of time.





Coherence



 Coherence - ratio of the cross spectral density to the power spectral density of the two signals

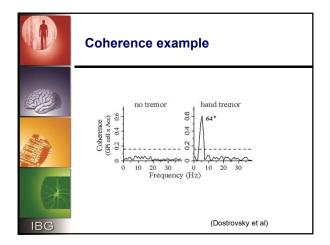


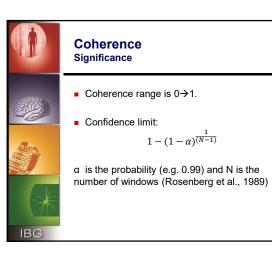
$$C_{x,y}(f) = \frac{\left|S_{x,y}(f)\right|^2}{S_{x,y}(f) \cdot S_{y,y}(f)}$$

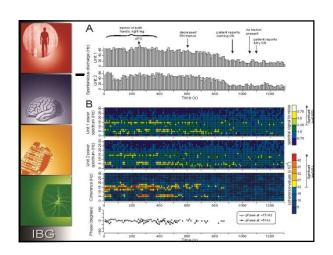


 Normalizes to the spectrum of the two signals and thus relates only to the relation between the signals and not to their structure.

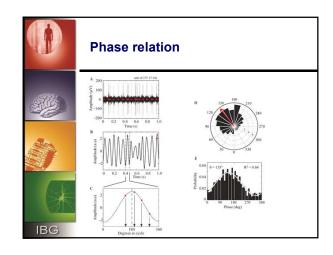
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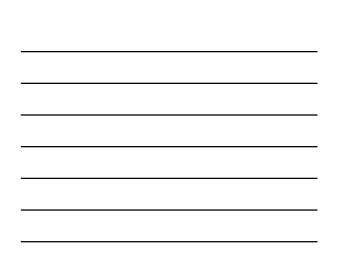


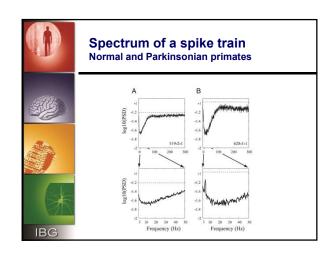


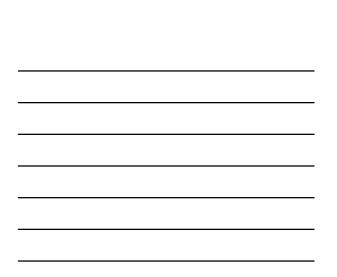


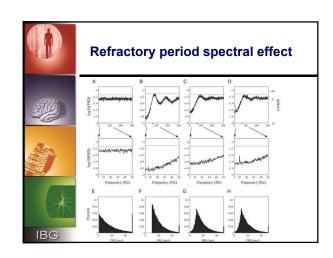




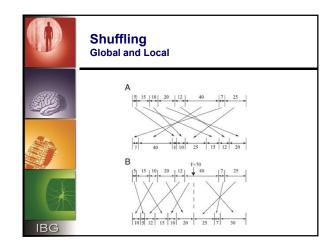


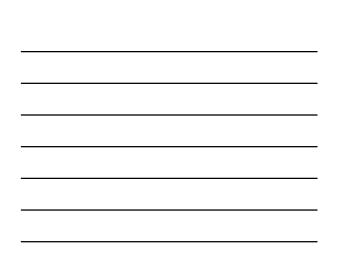


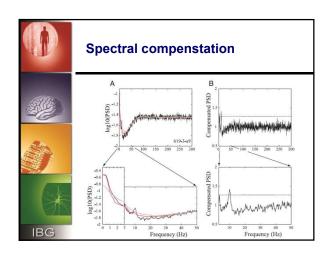


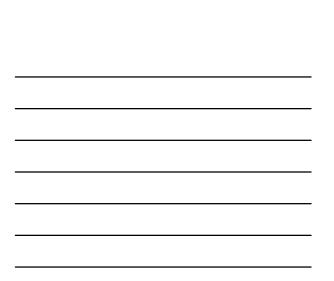


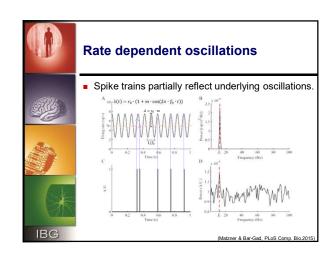


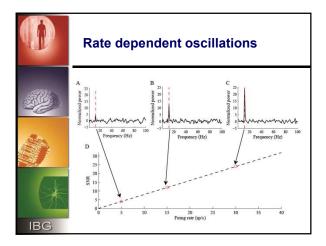








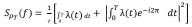






Power spectrum

■ An inhomogeneous spike train over a period (*T*):



Using the simple case of cosine rate modulation over a base frequency (f_0) : $S_{\rho_T}(f) = r_0 + r_0 m \operatorname{sinc}(2 f_0 T)$

 $+r_0^2T\mathrm{sinc}^2(fT)$

- $$\begin{split} &+r_0 \cdot T \sin(c(fT)) \sin((f-f_0)T) + r_0^2 T m \sin(fT) \sin((f+f_0)T) \\ &+ r_0^2 T m^2 \sin(fT) \sin((f-f_0)T) + \frac{r_0^2 T m^2}{4} \sin(fT) \sin((f+f_0)T) \\ &+ \frac{r_0^2 T m^2}{2} \sin((f-f_0)T) \sin((f+f_0)T) \end{split}$$

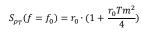


Spectral peak - function of the rate



The peak power $(f = f_0)$ is





 $S_{\rho_T}(f\neq f_0)=r_0$



The baseline power ($f \neq f_0$) is





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